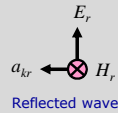


### Reflected wave

- It's traveling along  $-z$  axis

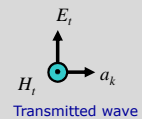


$$\vec{E}_{rs}(z) = E_{ro} e^{\gamma_1 z} \hat{x}$$

$$\vec{H}_{rs}(z) = H_{ro} e^{\gamma_1 z} (-\hat{y}) = -\frac{E_{ro}}{\eta_1} e^{\gamma_1 z} \hat{y}$$

### Transmitted wave

$$\vec{E}_{ts}(z) = E_{to} e^{-\gamma_2 z} \hat{x}$$



$$\vec{H}_{ts}(z) = H_{to} e^{-\gamma_2 z} \hat{y} = \frac{E_{to}}{\eta_2} e^{-\gamma_2 z} \hat{y}$$

### The total fields

- At medium 1 and medium 2

$$\vec{E}_1 = \vec{E}_i + \vec{E}_r \quad \vec{E}_2 = \vec{E}_t$$

$$\vec{H}_1 = \vec{H}_i + \vec{H}_r \quad \vec{H}_2 = \vec{H}_t$$

- Tangential components must be continuous at the interface

$$\vec{E}_i(0) + \vec{E}_r(0) = \vec{E}_t(0)$$

$$\vec{H}_i(0) + \vec{H}_r(0) = \vec{H}_t(0)$$

### Define

- Reflection coefficient,  $\Gamma$

$$\Gamma = \frac{E_{ro}}{E_{io}} = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1}$$

- Transmission coefficient,  $\tau$

$$\tau = \frac{E_{to}}{E_{io}} = \frac{2\eta_2}{\eta_2 + \eta_1}$$

Note:

- $1 + \Gamma = \tau$
- Both are dimensionless and may be complex
- $0 \leq |\Gamma| \leq 1$

### PE 10.8

A 5GHz uniform plane wave  $E_{is} = 10e^{-j\beta z} \hat{a}_x$  in free space is incident normally on a large plane, lossless dielectric slab ( $z > 0$ ) having  $\epsilon = 4\epsilon_0$  and  $\mu = \mu_0$ .

Find:

- the reflected wave  $E_{rs}$  and
- the transmitted wave  $E_{ts}$ .

### Case 1:

- Medium 1 = perfect dielectric

$$\sigma_1 = 0$$

- Medium 2 = perfect conductor

$$\sigma_2 = \infty$$

Halla impedancias int.

Reflección,

Transmisión

Y campos

$$\eta_2 = 0,$$

$$\Gamma = -1, \tau = 0$$

$$E_{1s} = -2jE_{io} \sin \beta_1 z \hat{x}$$

$$E_1 = 2E_{io} \sin \beta_1 z \sin \omega t \hat{x}$$

<http://www.phy.ntnu.edu.tw/java/waveSuperposition/waveSuperposition.html>

**The EM field forms a Standing Wave on medium 1**

$$E_1 = 2E_{io} \sin \beta_1 z \sin \omega t \hat{x}$$

Minima @  $-\beta_1 z = 0, \pi, 2\pi$   
 Maxima @  $-\beta_1 z = \frac{\pi}{2}, \frac{3\pi}{2}, \frac{5\pi}{2}$   
 $z_{\max} = -\frac{n\pi}{\beta_1} = -\frac{n\lambda_1}{4} \quad n = 1, 3, 5$

Conducting material  $\sigma_2 = \infty$

**Standing Wave Applets**

- ♦ <http://www.phy.ntnu.edu.tw/java/waveSuperposition/waveSuperposition.html>
- ♦ <http://www.ngsir.netfirms.com/englishhtm/StatWave.htm>
- ♦ <http://www.physics.smu.edu/~olness/www/03fall1320/applet/pipe-waves.html>

**Case 2:**

- ♦ Medium 1 = perfect dielectric  $\sigma_1=0$
- ♦ Medium 2 = perfect dielectric  $\sigma_2=0$

If  $\eta_2 > \eta_1$ ,  
 $\Gamma > 0$ ,  
 $\tau$  and  $\Gamma$  are real.

$$E_{1s} = E_{is} + E_{rs} = E_{oi}(e^{-j\beta_1 z} + \Gamma e^{+j\beta_1 z}) = E_{oi}e^{-j\beta_1 z}(1 + \Gamma e^{+2j\beta_1 z})$$

$-2\beta_1 z_{\max} = 0, 2\pi, 4\pi, 6\pi, \dots$   
 or  $-\beta_1 z_{\max} = 0, \pi, 2\pi, 3\pi, \dots$

$$z_{\max} = -\frac{n\pi}{\beta_1} = -\frac{n\lambda_1}{2} \quad n = 0, 1, 2, 3$$

$$z_{\min} = -\frac{(2n+1)\pi}{2\beta_1} = -\frac{(2n+1)\lambda_1}{4} \quad n = 0, 1, 2, 3$$

**Standing waves due to reflection**

$$E_1 = E_i + E_r = E_{oi}(e^{-j\beta_1 z} + \Gamma e^{+j\beta_1 z}) = E_{oi}e^{-j\beta_1 z}(1 + \Gamma e^{+2j\beta_1 z})$$

Lossless Medium 1  $\sigma_1=0$   
 $\eta_2 > \eta_1$

$z_{\max} = -\frac{n\pi}{\beta_1} = -\frac{n\lambda_1}{2} \quad n = 0, 1, 2, 3$

Lossless Medium 2  $\sigma_2 = 0$

**Case 3:**

- ♦ Medium 1 = perfect dielectric  $\sigma_1=0$
- ♦ Medium 2 = perfect dielectric  $\sigma_2=0$

If  $\eta_2 < \eta_1$ ,  
 $\Gamma < 0$ ,  $\tau$  and  $\Gamma$  are real.

$$z_{\max} = -\frac{(2n+1)\pi}{2\beta_1} = -\frac{(2n+1)\lambda_1}{4} \quad n = 1, 2, 3$$

$$z_{\min} = -\frac{n\pi}{\beta_1} = -\frac{n\lambda_1}{2} \quad n = 0, 1, 2, 3$$

**Standing waves due to reflection**

$$E_1 = E_i + E_r = E_{oi}(e^{-j\beta_1 z} + \Gamma e^{+j\beta_1 z}) = E_{oi}e^{-j\beta_1 z}(1 + \Gamma e^{+2j\beta_1 z})$$

Lossless Medium 1  $\eta_2 < \eta_1$

$z_{\max} = -\frac{(2n+1)\lambda_1}{4} \quad n = 0, 1, 2, 3$

Lossless Medium 2  $\sigma_2 = 0$

## Standing Wave Ratio, $s$

- Measures the amount of reflections, the more reflections, the larger the standing wave that is formed.
- The ratio of  $|E_1|_{\max}$  to  $|E_1|_{\min}$

$$s = \frac{|E_1|_{\max}}{|E_1|_{\min}} = \frac{|H_1|_{\max}}{|H_1|_{\min}} = \frac{1+|\Gamma|}{1-|\Gamma|}$$

or

$$|\Gamma| = \frac{s-1}{s+1}$$

Ideally  
(0dB)  
No reflections

## PE 10.9

- The plane wave  $E = 50 \sin(\omega t - 5x) \mathbf{a}_y$  V/m in a lossless medium ( $\mu = 4\mu_0$ ,  $\epsilon = \epsilon_0$ ) encounters a lossy medium ( $\mu = \mu_0$ ,  $\epsilon = 4\epsilon_0$ ,  $\sigma = 0.1$  mhos/m) normal to the  $x$ -axis at  $x=0$ . Find

- $\Gamma$
- $\tau$
- $s$
- $E_r$
- $E_t$

## Ex. Antenna Radome



A 10GHz aircraft radar uses a narrow-beam scanning antenna mounted on a gimbal behind a dielectric radome.

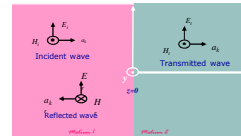
- Even though the radome shape is far from planar, it is approximately planar over the narrow extent of the radar beam.
- If the radome material is a lossless dielectric with  $\mu_r=1$  and  $\epsilon_r=9$ , choose its thickness  $d$  such that the radome appears transparent to the radar beam.
- Mechanical integrity requires  $d$  to be greater than 2.3 cm.



## Power Flow in Medium 1

- The net average power density flowing in medium 1

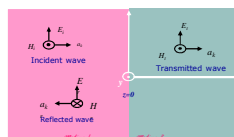
$$\begin{aligned} P_{ave1}(z) &= \frac{1}{2} \text{Re}[E_1 \times H_1^*] \\ &= \frac{1}{2} \text{Re} \left[ \hat{x} E_{io} (e^{-j\beta_1 z} + \Gamma e^{j\beta_1 z}) \times \hat{y} \frac{E_{io}^*}{\eta_1} (e^{j\beta_1 z} + \Gamma^* e^{-j\beta_1 z}) \right] \\ &= \hat{z} \frac{|E_{io}|^2}{2\eta_1} (1 - |\Gamma|^2) \\ &= P_{ave}^i + P_{ave}^r \end{aligned}$$



## Power Flow in wave

- The net average power density flowing in medium 2

$$\begin{aligned} P_{ave2}(z) &= \frac{1}{2} \text{Re}[E_2 \times H_2^*] \\ &= \frac{1}{2} \text{Re} \left[ \hat{x} \tau E_{io} e^{-j\beta_2 z} \times \hat{y} \tau^* \frac{E_{io}^*}{\eta_2} e^{j\beta_2 z} \right] \\ &= \hat{z} |\tau|^2 \frac{|E_{io}|^2}{2\eta_2} \\ \text{where } \hat{a}_k &= \hat{z} \end{aligned}$$



## Power in Media

$$\begin{aligned} P_{ave1}(z) &= \hat{z} \frac{|E_o|^2}{2\eta_{c1}} (e^{-2\alpha_1 z} - |\Gamma|^2 e^{2\alpha_1 z}) \\ P_{ave2}(z) &= \hat{z} |\tau|^2 \frac{|E_o|^2}{2} e^{-2\alpha_2 z} \text{Re} \left( \frac{1}{\eta_{c2}^*} \right) \end{aligned}$$

where

$$\Gamma = \frac{\sqrt{\epsilon_{c1}} - \sqrt{\epsilon_{c2}}}{\sqrt{\epsilon_{c1}} + \sqrt{\epsilon_{c2}}} \quad \text{and} \quad \epsilon_{c2} = \epsilon_2 - j \frac{\sigma_2}{\omega_2}$$

## We will look at...

### I. Normal incidence

Wave arrives at  $90^\circ$  from the surface

Standing waves



### II. Oblique incidence ( )

Wave arrives at an angle

Snell's Law and Critical angle

Parallel or Perpendicular

Brewster angle

