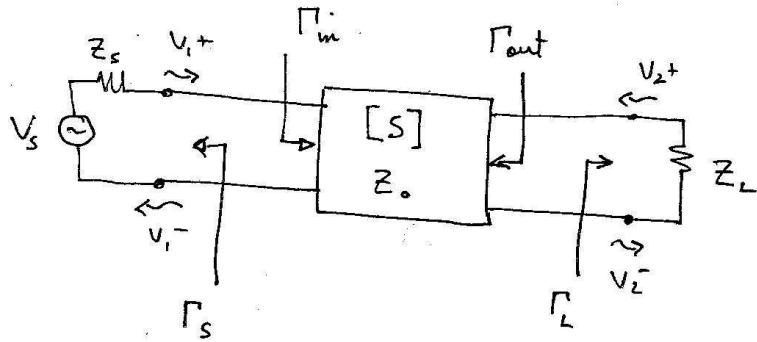


## Transistor Amplifier - Power Gain Eqs.



Usando como referencia  $Z_0$ ,  $\Gamma_s = \frac{Z_s - Z_0}{Z_s + Z_0}$ ,  $\Gamma_L = \frac{Z_L - Z_0}{Z_L + Z_0}$

Sabemos que:  $V_i^- = S_{11}V_i^+ + S_{12}V_o^+$ ,  $V_o^- = S_{21}V_i^+ + S_{22}V_o^+$

$$V_o^- = S_{21}V_i^+ + S_{22}\Gamma_L V_o^- \quad \text{pero}$$

$$\Gamma_{in} = \frac{V_i^-}{V_i^+}$$

De aquí,

$$V_i^- = S_{11}V_i^+ + S_{12}\Gamma_L V_o^- \quad (*)$$

$$\Gamma_L = \frac{V_o^+}{V_o^-}$$

$$\boxed{V_o^- = S_{21}V_i^+ + S_{22}\Gamma_L V_o^-}$$

$$\rightarrow V_o^- = \frac{S_{21}V_i^+}{1 - S_{22}\Gamma_L}$$

Sustituyendo en (\*)

de aquí

$$V_i^- = S_{11}V_i^+ + S_{12}\Gamma_L \left( \frac{S_{21}V_i^+}{1 - S_{22}\Gamma_L} \right) \quad \text{Eq. en términos de } V_i^+ \text{ y } V_i^-$$

$$\Gamma_{in} = \frac{V_i^-}{V_i^+} = S_{11} + \frac{S_{12}S_{21}\Gamma_L}{1 - S_{22}\Gamma_L} = \frac{\Gamma_{in} - Z_0}{\Gamma_{in} + Z_0} \quad (1)$$

De la misma forma,

$$\Gamma_{out} = \frac{V_o^+}{V_o^-} = S_{22} + \frac{S_{12}S_{21}\Gamma_s}{1 - S_{11}\Gamma_s} \quad (2)$$

Deseo expresión de potencia que entra a la red.

$$P_m = \frac{|V_i|^2}{2Z_0} (1 - |\Gamma_m|^2)$$

Hallar expresión para  $V_i^+$ :

$$V_i = \frac{V_s Z_m}{Z_s + Z_m} = V_i^+ + V_i^- = V_i^+ (1 + \Gamma_m)$$

$$V_i^+ = \frac{V_s Z_m}{(Z_s + Z_m)(1 + \Gamma_m)}, \text{ pero } Z_m = Z_0 \frac{1 + \Gamma_m}{1 - \Gamma_m}$$

$$\therefore V_i^+ = \frac{V_s \cancel{\frac{1 + \Gamma_m}{1 - \Gamma_m}} Z_0}{(Z_s + Z_m) \cancel{(1 + \Gamma_m)}} = \frac{V_s Z_0}{(Z_s + Z_0)(1 - \Gamma_m)}$$

$$V_i^+ = \frac{V_s Z_0}{Z_0 \left( \frac{1 + \Gamma_s}{1 - \Gamma_s} + \frac{1 + \Gamma_m}{1 - \Gamma_m} \right) (1 - \Gamma_m)} = \frac{V_s (1 - \Gamma_s)}{2 (1 - \Gamma_s \Gamma_m)}$$

$$P_m = \frac{|V_s|^2 |1 - \Gamma_s|^2}{8 Z_0 |1 - \Gamma_s \Gamma_m|^2} (1 - |\Gamma_m|^2)$$

De la misma forma,

$$P_L = \frac{|V_z|^2}{2Z_0} (1 - |\Gamma_L|^2)$$

$$P_L = \frac{|V_s|^2 |S_{21}|^2 (1 - |\Gamma_L|^2) |1 - \Gamma_s|^2}{8 Z_0 |1 - S_{22} \Gamma_L|^2 |1 - \Gamma_s \Gamma_m|^2}$$

Definimos "Power Gain"

3

$$G = \frac{P_L}{P_{in}} = \frac{|S_{21}|^2 (1 - |\Gamma_L|^2)}{(1 - |\Gamma_{in}|^2) |1 - S_{22} \Gamma_L|^2}$$

Depende de  
 $\Gamma_L$

Eq. 11.8

"Available Gain",  $G_A$

$$G_A = \frac{P_{AVN}}{P_{AVS}} ; \quad P_{AVN} = P_L \left| \begin{array}{l} \Gamma_L = \Gamma_{out}^* \end{array} \right.$$

$$P_{AVS} = P_{in} \left| \begin{array}{l} \Gamma_{in} = \Gamma_S^* \end{array} \right.$$

$$P_{AVS} = \frac{|V_s|^2 |1 - \Gamma_S|^2}{8Z_0 |1 - |\Gamma_S|^2|^2} \left( \frac{1}{1 - |\Gamma_S|^2} \right) = \frac{|V_s|^2 |1 - \Gamma_S|^2}{8Z_0 |1 - |\Gamma_S|^2|^2}$$

$$P_{AVN} = \frac{|V_s|^2 |S_{21}|^2 (1 - |\Gamma_{out}|^2) |1 - \Gamma_S|^2}{8Z_0 |1 - S_{22} \Gamma_{out}^*|^2 |1 - \Gamma_S \Gamma_{in}|^2}$$

pero  $\Gamma_{in} = S_{11} + \frac{S_{12} S_{21} \Gamma_L}{1 - S_{22} \Gamma_L^*} = S_{11} + \frac{S_{12} S_{21} \Gamma_{out}^*}{1 - S_{22} \Gamma_{out}^*}$

$$P_{AVN} = \frac{|V_s|^2 |S_{21}|^2 |1 - \Gamma_S|^2}{8Z_0 |(1 - S_{11} \Gamma_S)^2 (1 - |\Gamma_{out}|^2)|}$$

$$G_A = \frac{P_{AVN}}{P_{AVS}} = \frac{|S_{21}|^2 (1 - |\Gamma_S|^2)}{|1 - S_{11} \Gamma_S|^2 (1 - |\Gamma_{out}|^2)}$$

Depende de  
 $\Gamma_S$   
Eq. 11.12

Transistor Power Gain

$$G_T = \frac{P_L}{P_{\text{AUS}}} = \frac{|S_{21}|^2 (1 - |\Gamma_s|^2) (1 - |\Gamma_L|^2)}{|1 - S_{11} \Gamma_s|^2 |1 - S_{22} \Gamma_L|^2}$$

4  
Depende  
 $\Gamma_s$  y  $\Gamma_L$   
Eq. 11.13

Si  $\Gamma_L = \Gamma_s = 0$        $G_T = |S_{21}|^2$

Caso especial  $\rightarrow$  caso unilateral  $\Rightarrow S_{12} = 0$

Si  $S_{12} = 0$  entonces  $\Gamma_{in} = S_{11}$  ,  $\Gamma_{out} = S_{22}$   
 y  $G_T$  se reduce a  $(\text{eq. 1})$   $(\text{eq. 2})$

$$G_{Tu} = \frac{|S_{21}|^2 (1 - |\Gamma_s|^2) (1 - |\Gamma_L|^2)}{|1 - S_{11} \Gamma_s|^2 |1 - S_{22} \Gamma_L|^2}$$

Ejemplo :  $S_{11} = 0.45 \angle 150^\circ$        $S_{12} = 0.01 \angle -10^\circ$   
 $S_{21} = 2.05 \angle 10^\circ$        $S_{22} = 0.4 \angle -150^\circ$

Si  $Z_s = 20\Omega$       y       $Z_L = 30\Omega$

$\Gamma_s = -0.429$  ;       $\Gamma_L = -0.250$

De aqui ;  $\Gamma_{in} = S_{11} + \frac{S_{12} S_{21} \Gamma_L}{1 - S_{22} \Gamma_L} = 0.455 \angle 150^\circ$

$\Gamma_{out} = 0.408 \angle -151^\circ$

$G = 5.94$        $G_A = 5.85$        $G_T = 5.49$

Fórmula Dividir ganancia en tres términos:

$$G_T = G_s \quad G_c \quad G_L$$

Donde  $G_s = \frac{1 - |\Gamma_s|^2}{|1 - S_{11} \Gamma_s|^2}$ ;  $G_c = |S_{21}|^2$ ;  $G_L = \frac{1 - |\Gamma_L|^2}{|1 - S_{22} \Gamma_L|^2}$

$G_s \Rightarrow$  Input Matching Network

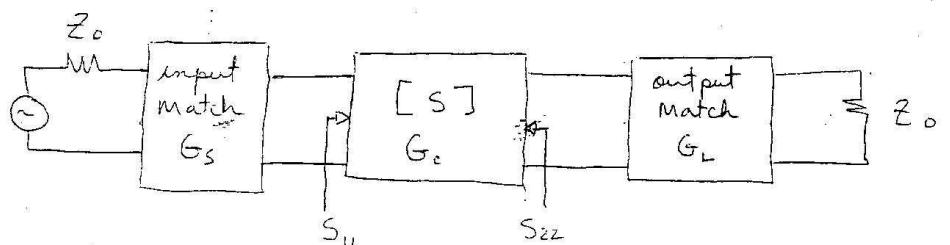
$G_c \Rightarrow$  output " "

$G_L \Rightarrow$  transistor ( $S_{21}$ )

Caso especial (unilateral) existe si  $[S_{12} = 0]$

De esta condición:  $\Gamma_m = S_{11}$ ;  $\Gamma_{out} = S_{22}$

$$G_{Tu} = \frac{|S_{21}|^2 (1 - |\Gamma_s|^2) (1 - |\Gamma_L|^2)}{|1 - S_{11} \Gamma_s|^2 |1 - S_{22} \Gamma_L|^2}$$



$$G_s = \frac{1 - |\Gamma_s|^2}{|1 - S_{11} \Gamma_s|^2} \quad G_c = |S_{21}|^2 \quad G_L = \frac{1 - |\Gamma_L|^2}{|1 - S_{22} \Gamma_L|^2}$$

Próximo Tema: Estabilidad

Estabilidad : El amplificador es estable

- si  $|\Gamma_{in}| < 1$  y  $|\Gamma_{out}| < 1$ .  $\Gamma_{in}$  y  $\Gamma_{out}$  dependen de  $\Gamma_s$  y  $\Gamma_L$ , por lo que  $\Gamma_s$  y  $\Gamma_L$  controlan la estabilidad del amplificador.

Deseamos valores de  $\Gamma_L$  y  $\Gamma_s$  para que  $\Gamma_{in}$  y  $\Gamma_{out}$  resulten  $< 1$ . Habrá valores

que ocasionen :  $\left. \begin{array}{l} |\Gamma_{in}| > 1 \\ |\Gamma_{out}| > 1 \end{array} \right\} \text{inestables}$

Usando las ecuaciones de  $\Gamma_{in}$  y  $\Gamma_{out}$ , podemos derivar expresiones para identificar valores de  $\Gamma_L$  y  $\Gamma_s$  deseados; usando el "Smith Chart". Hay dos tipos :

1) Estabilidad sin condición :  $|\Gamma_{in}|$  y  $|\Gamma_{out}|$  son  $< 1$  para cualquier  $\Gamma_L$  y  $\Gamma_s$ .

2) Estabilidad condicional  $|\Gamma_{in}|$  y  $|\Gamma_{out}| < 1$  para algunos valores de  $\Gamma_L$  y  $\Gamma_s$ . ¿Cómo identificar estos valores?

4

$|\Gamma_m| = 1 \quad |\Gamma_{out}| = 1$  podemos  
determinar valores de  $\Gamma_s$  y  $\Gamma_L$  donde ésto  
se cumple.

$$|\Gamma_m| = \left| S_{11} + \frac{S_{12} S_{21} \Gamma_L}{1 - S_{22} \Gamma_L} \right| = 1$$

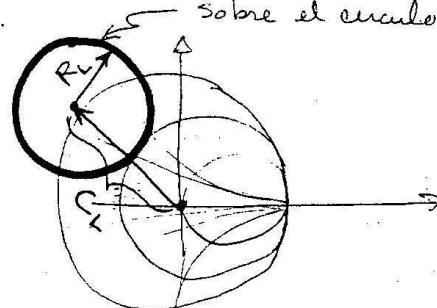
$$|\Gamma_{out}| = \left| S_{22} + \frac{S_{12} S_{21} \Gamma_s}{1 - S_{11} \Gamma_s} \right| = 1$$

De aquí resulta ecuaciones de círculo:

Centro:  $C_L = \frac{(S_{22} - \Delta S_{11})^*}{|S_{22}|^2 - |\Delta|^2}$  Radio:  $R_L = \left| \frac{S_{12} S_{21}}{|S_{22}|^2 - |\Delta|^2} \right|$

En plano de  $\Gamma_L$ : sobre el círculo:  $|\Gamma_m| = 1$

"Output Stability  
circle"

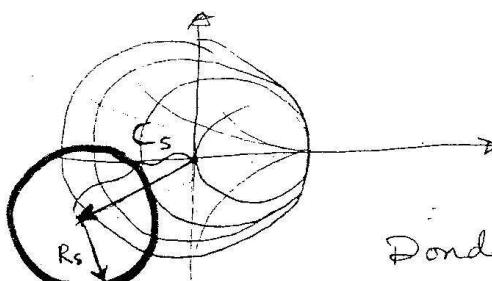


En plano de  $\Gamma_s$

"Input Stability  
circle"

$$C_S = \frac{(S_{11} - \Delta S_{22})^*}{|S_{11}|^2 - |\Delta|^2}$$

$$R_S = \left| \frac{S_{12} S_{21}}{|S_{11}|^2 - |\Delta|^2} \right|$$



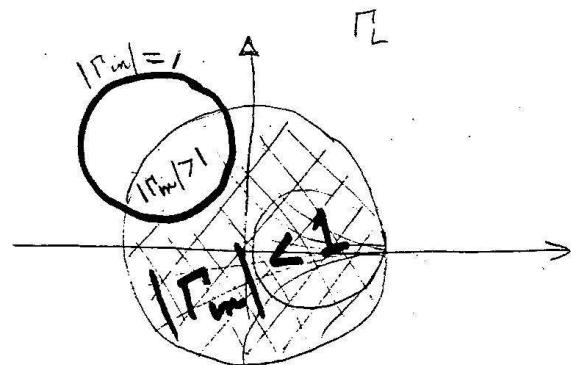
Donde:  
 $\Delta = S_{11} S_{22} - S_{12} S_{21}$

Como determinar si la región estable es dentro o fuera del círculo?

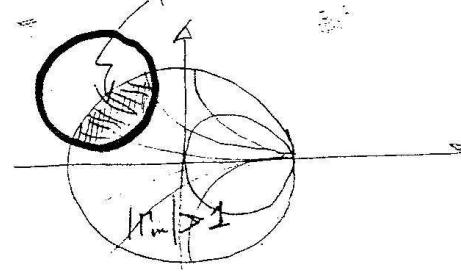
Un punto fácil:  $|R_m| = 0$  (centro del S.C.)

$$\therefore |R_m| = |S_{11} + \frac{S_{12}S_{21}R_L}{1 - S_{22}R_L}| = |S_{11}|$$

Si  $|S_{11}| < 1$ , la región que contiene el centro del S.C. es estable

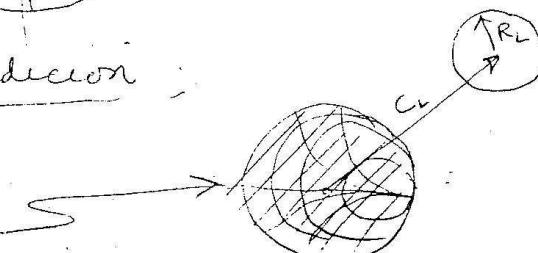


Si  $|S_{11}| > 1$ ,  $|R_m| < 1$  (Región estable)



Estable sin condición

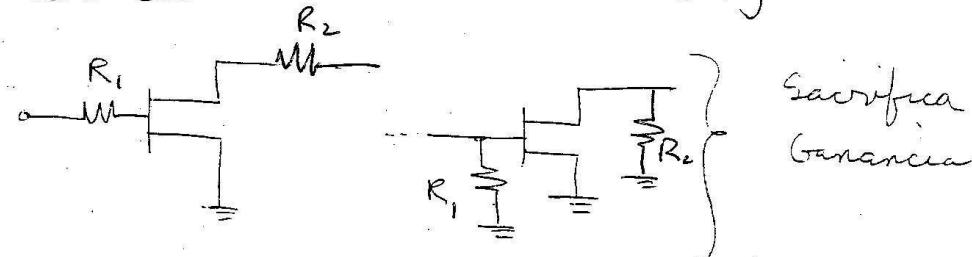
Todo  $R_L$   
resulta en  
 $|R_m| < 1$



si el transistor es estable con conducción,

podemos cambiar esta condición a estable

sin conducción con "resistive loading".



Usar uno de los dos ( $R_1, R_2$ )

### Diseño de Amplificador (Garantía máxima)

$$\Gamma_{in} = \Gamma_s^* \quad (\text{maxima transferencia en entrada de amplificador})$$

$$\Gamma_{out} = \Gamma_L^* \quad (\text{maxima transferencia en salida del amplificador})$$

$$G_{T\max} = \frac{1}{1 - |\Gamma_s|^2} \left( S_{21} \frac{f^2}{P_s} \right) \frac{1 - |\Gamma_L|^2}{|1 - S_{22} \Gamma_L|^2}$$

$$\Gamma_{in} = \Gamma_s^* = S_{11} + \frac{S_{12} \Gamma_L S_{21}}{1 - S_{22} \Gamma_L}$$

Dos errores  
dos desono-  
cadas:

$$\Gamma_L \text{ y } \Gamma_s$$

$$\Gamma_{out} = \Gamma_L^* = S_{22} + \frac{S_{12} S_{21} \Gamma_s}{1 - S_{11} \Gamma_s}$$

De aquí resulta que:

$$\Gamma_s = \frac{B_1 \pm \sqrt{B_1^2 - 4|C_1|^2}}{2C_1}$$

$$\Gamma_L = \frac{B_2 \pm \sqrt{B_2^2 - 4|C_2|^2}}{2C_2}$$

Donde  $B_1 = 1 + |S_{11}|^2 - |S_{22}|^2 - |\Delta|^2$

$$B_2 = 1 + |S_{22}|^2 - |S_{11}|^2 - |\Delta|^2$$

$$C_1 = S_{11} - \Delta S_{22}^*$$

$$C_2 = S_{22} - \Delta S_{11}^*$$

Para caso unilateral;

$$\Gamma_s = S_{11}^* \quad \Gamma_L = S_{22}^*$$

$$G_{\text{ru max}} = \frac{1}{1 - |S_{11}|^2} |S_{21}|^2 \frac{1}{1 - |S_{22}|^2}$$

Otras condiciones matemáticas de estabilidad

Estable sin condición:

$$K = \frac{1 - |S_{11}|^2 - |S_{22}|^2 + |\Delta|^2}{2|S_{12}S_{21}|} > 1$$

$$\text{y } |\Delta| < 1$$

$$M = \frac{1 - |S_{11}|^2}{|S_{22} - S_{11}^* \Delta| + |S_{21}S_{12}|} > 1$$