



Wave Incidence

[Chapter 10 cont, Sadiku]

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Ex. Light traveling in air encounters the water; another medium.



Wave incidence

- For many applications, [such as fiber optics, line power transmission], it's necessary to know what happens to a wave when it meets a different medium.
 - How much is *transmitted*?
 - How much is *reflected* back?

SEE Java Applets
<http://physics.usask.ca/~hirose/ep225/anim.htm>

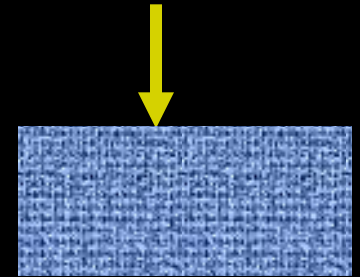


We will look at...

I. Normal incidence

Wave arrives at 0° from normal

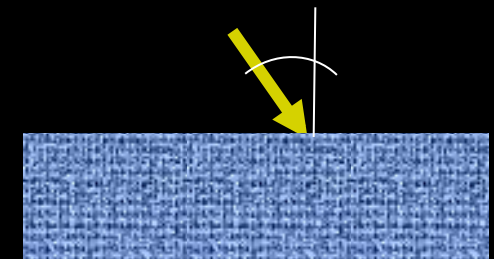
- “Standing waves”



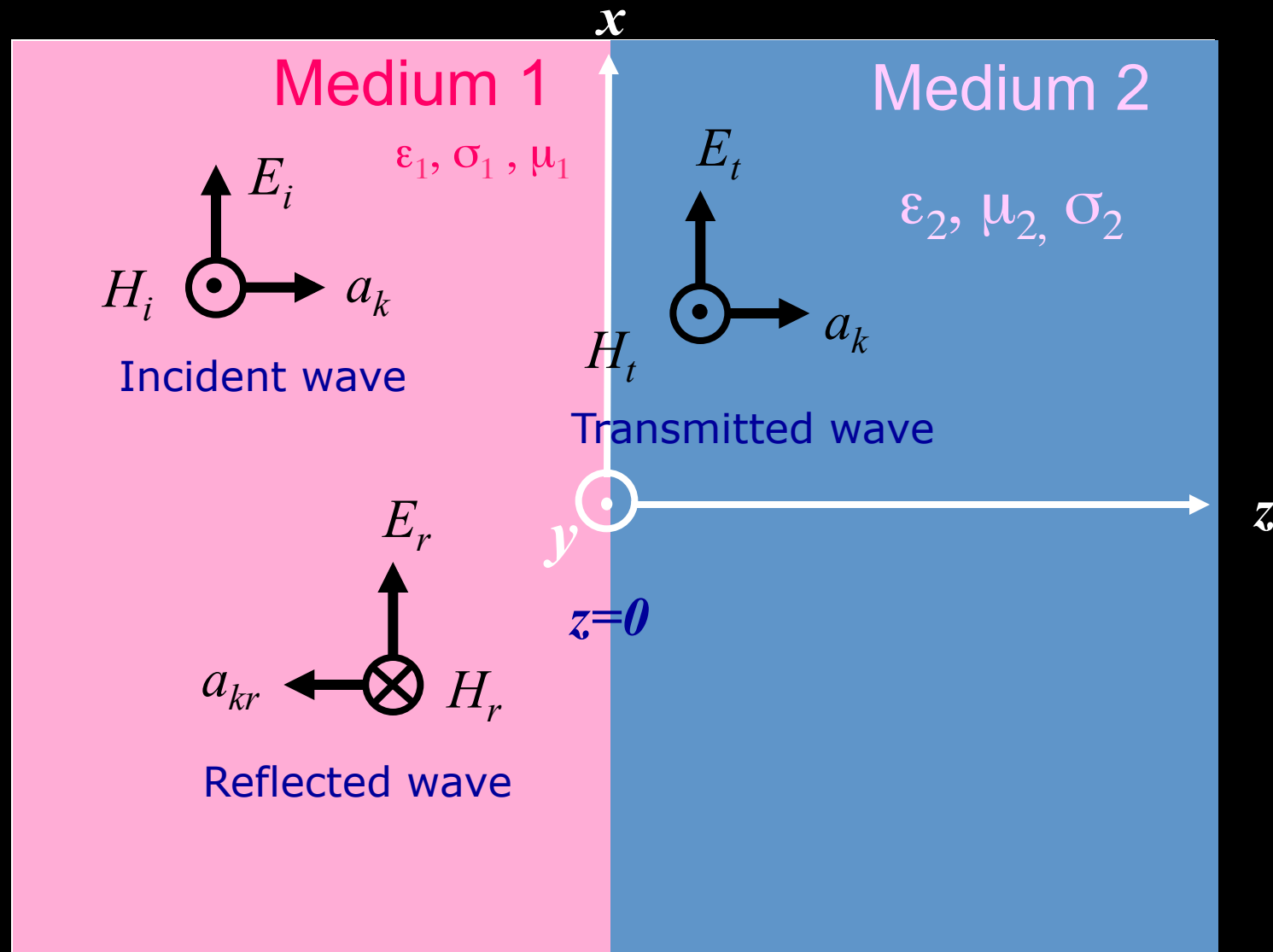
II. Oblique incidence

Wave arrives at another angle

- Snell's Law and Critical angle
- Parallel or Perpendicular polarization
- Brewster angle

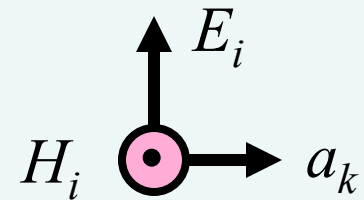


Reflection at Normal Incidence



Now in terms of equations ...

- Incident wave



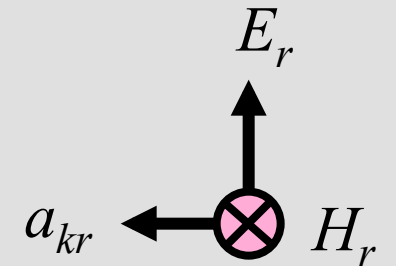
Incident wave

$$\vec{E}_{is}(z) = E_{io} e^{-\gamma_1 z} \hat{x}$$

$$\vec{H}_{is}(z) = H_{io} e^{-\gamma_1 z} \hat{y} = \frac{E_{io}}{\eta_1} e^{-\gamma_1 z} \hat{y}$$

Reflected wave

- It's traveling along $-z$ axis



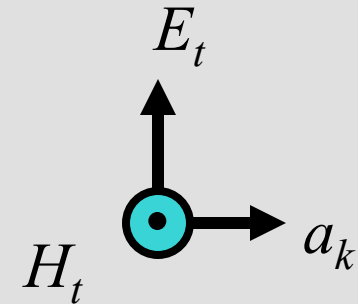
Reflected wave

$$\vec{E}_{rs}(z) = E_{ro} e^{\gamma_1 z} \hat{x}$$

$$\vec{H}_{rs}(z) = H_{ro} e^{\gamma_1 z} (-\hat{y}) = -\frac{E_{ro}}{\eta_1} e^{\gamma_1 z} \hat{y}$$

Transmitted wave

$$\vec{E}_{ts}(z) = E_{to} e^{-\gamma_2 z} \hat{x}$$



Transmitted wave

$$\vec{H}_{ts}(z) = H_{to} e^{-\gamma_2 z} \hat{y} = \frac{E_{to}}{\eta_2} e^{-\gamma_2 z} \hat{y}$$

The total fields

- At medium 1 and medium 2

$$\begin{aligned}\vec{E}_1 &= \vec{E}_i + \vec{E}_r & \vec{E}_2 &= \vec{E}_t \\ \vec{H}_1 &= \vec{H}_i + \vec{H}_r & \vec{H}_2 &= \vec{H}_t\end{aligned}$$



- Tangential components must be continuous at the interface

$$\begin{aligned}\vec{E}_i(0) + \vec{E}_r(0) &= \vec{E}_t(0) \\ \vec{H}_i(0) + \vec{H}_r(0) &= \vec{H}_t(0)\end{aligned}$$

Define

- Reflection coefficient, Γ

$$\Gamma = \frac{E_{ro}}{E_{io}} = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1}$$

- Transmission coefficient, τ

$$\tau = \frac{E_{to}}{E_{io}} = \frac{2\eta_2}{\eta_2 + \eta_1}$$

Note:

- $1 + \Gamma = \tau$
- Both are dimensionless and may be complex
- $0 \leq |\Gamma| \leq 1$

PE 10.8

A 5GHz uniform plane wave $E_{is} = 10e^{-j\beta z} a_x$ in free space is incident normally on a large plane, lossless dielectric slab ($z > 0$) having $\epsilon = 4\epsilon_0$ and $\mu = \mu_0$.

Find:

- ◆ the reflected wave E_{rs} and
- ◆ the transmitted wave E_{ts} .

SEE <http://www.acs.psu.edu/drussell/Demos/reflect/reflect.html>

Answer:

$$-3.33 e^{j\beta_1 z} a_x \text{ V/m,}$$

$$6.67 e^{-j\beta_2 z} a_x \text{ V/m}$$

$$\text{where } \beta_2 = 2\beta_1 = 200 \pi/3$$

Case 1:

- Medium 1: perfect dielectric, $\sigma_1=0$
- Medium 2: perfect conductor, $\sigma_2=\infty$

Halla impedancias intrínseca.

Reflección,
Transmisión
y campos

$$(1 + j) \frac{\alpha}{\sigma}$$

$$\eta_2 = 0,$$

$$\Gamma = -1, \tau = 0$$

$$E_{1s} = -2jE_{io} \sin \beta_1 z \hat{x} \text{ (phasor)}$$

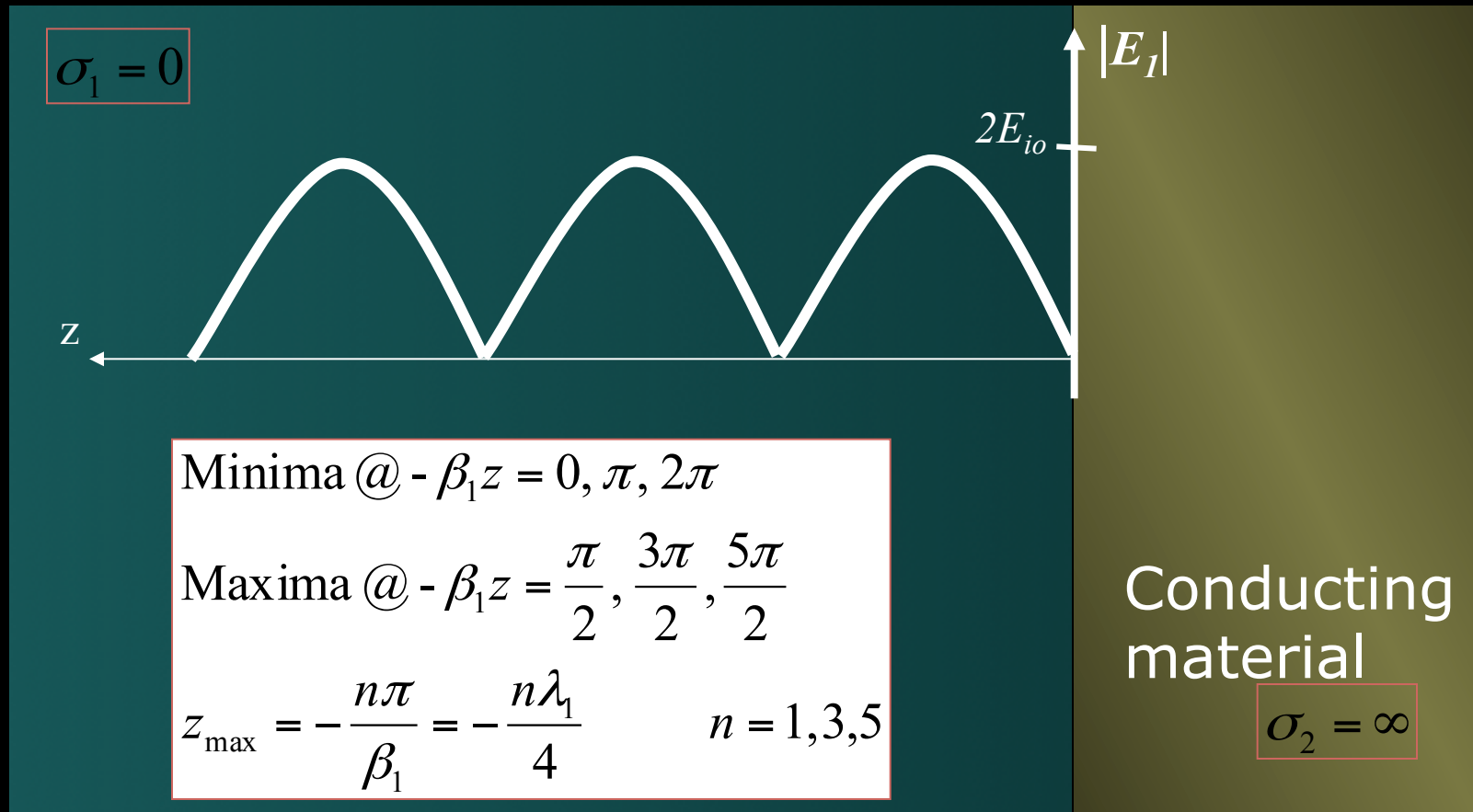
$$E_1(z, t) = 2E_{io} \sin \beta_1 z \sin \omega t \hat{x}$$

<http://www.phy.ntnu.edu.tw/java/waveSuperposition/waveSuperposition.html>



The EM field forms a Standing Wave on medium 1

$$E_1 = 2E_{i0} \sin \beta_1 z \sin \omega t \hat{x}$$



Standing Wave Applets

- <http://www.walter-fendt.de/ph14e/stwaverefl.htm>

Case 2:

- ◆ Medium 1: perfect dielectric $\sigma_1=0$
- ◆ Medium 2: perfect dielectric $\sigma_2=0$

If $\eta_2 > \eta_1$,

$\Gamma > 0$,

τ and Γ are real.

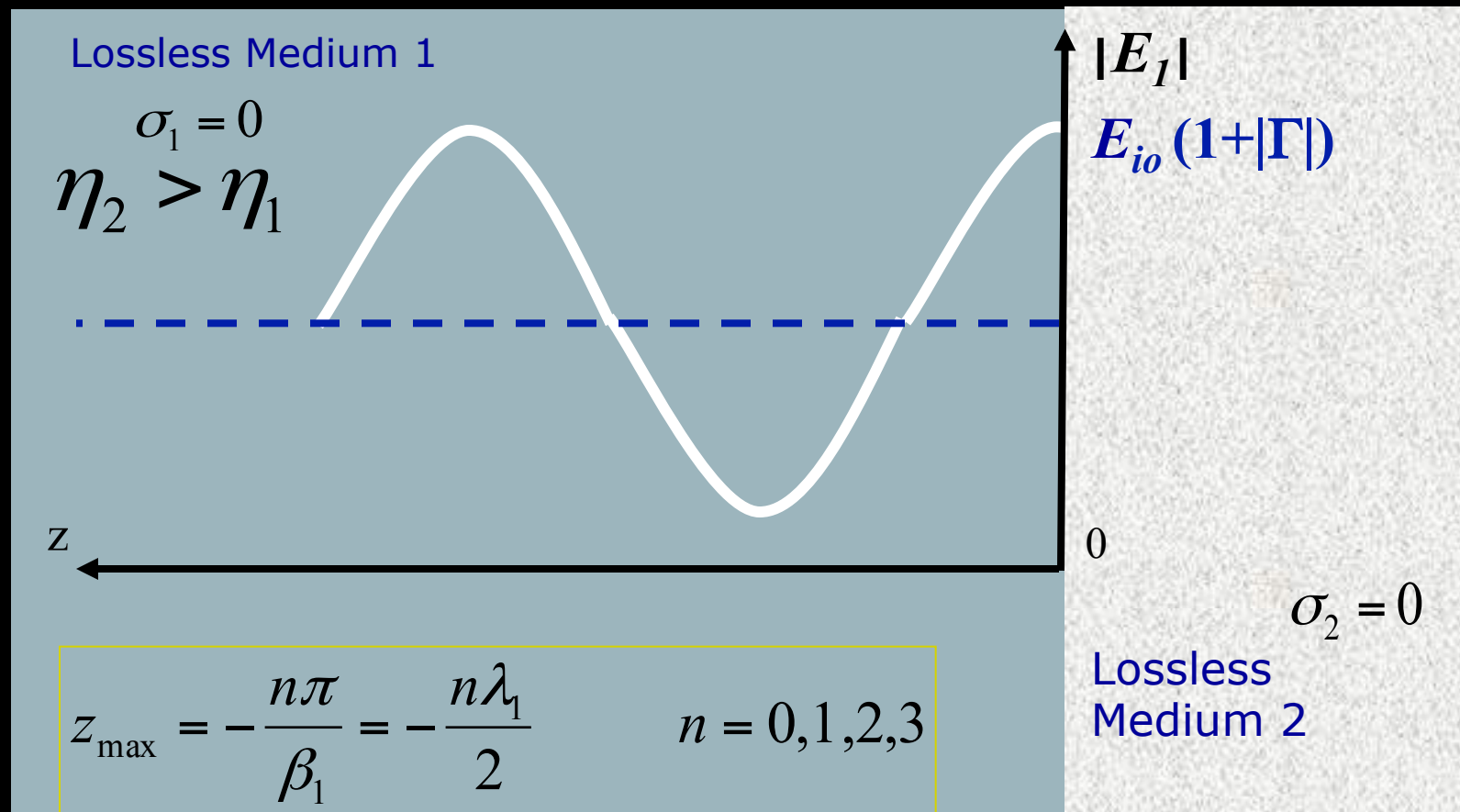
$$\begin{aligned} E_{1s} &= E_{is} + E_{rs} \\ &= E_{oi} (e^{-j\beta_1 z} + \Gamma e^{+j\beta_1 z}) \\ &= E_{oi} e^{-j\beta_1 z} (1 + \Gamma e^{+2j\beta_1 z}) \end{aligned}$$

$$\begin{aligned} -2\beta_1 z_{\max} &= 0, 2\pi, 4\pi, 6\pi, \dots \\ \text{or } -\beta_1 z_{\max} &= 0, \pi, 2\pi, 3\pi, \dots \end{aligned}$$

$$z_{\max} = -\frac{n\pi}{\beta_1} = -\frac{n\lambda_1}{2} \quad n = 0, 1, 2, 3$$

Standing waves due to reflection

$$E_1 = E_i + E_r = E_{oi} (e^{-j\beta_1 z} + \Gamma e^{+j\beta_1 z}) = E_{oi} e^{-j\beta_1 z} (1 + \Gamma e^{+2j\beta_1 z})$$



***** At every half-wavelength, everything repeats! *****

Case 3:

- ◆ Medium 1 = perfect dielectric $\sigma_1=0$
- ◆ Medium 2 = perfect dielectric $\sigma_2=0$

If $\eta_2 < \eta_1$,

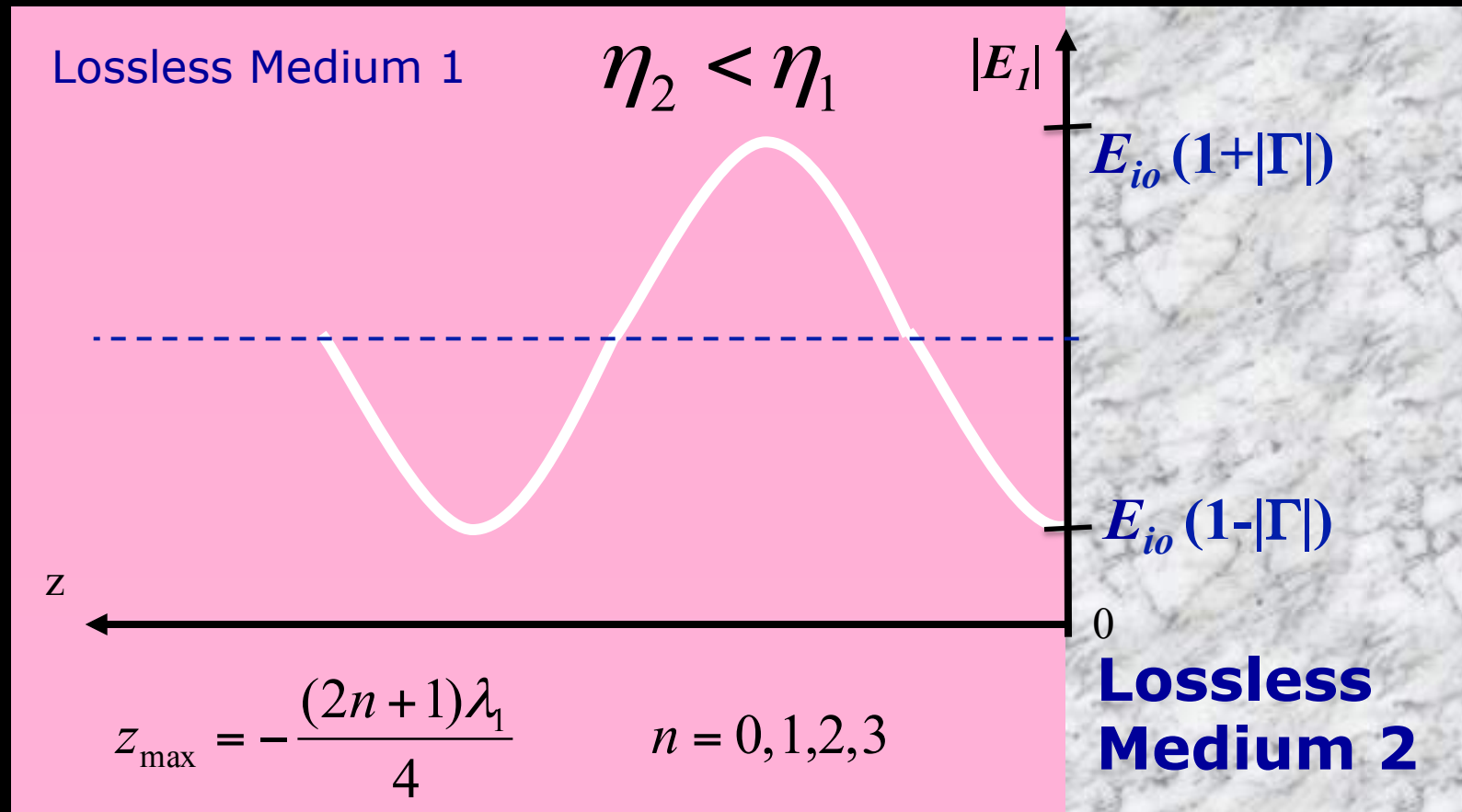
$\Gamma < 0$, τ and Γ are real.

$$Z_{\max} = -\frac{(2n+1)\pi}{2\beta_1} = -\frac{(2n+1)\lambda_1}{4} \quad n = 1, 2, 3$$

$$Z_{\min} = -\frac{n\pi}{\beta_1} = -\frac{n\lambda_1}{2} \quad n = 0, 1, 2, 3$$

Standing waves due to reflection

$$E_1 = E_i + E_r = E_{oi} (e^{-j\beta_1 z} + \Gamma e^{+j\beta_1 z}) = E_{oi} e^{-j\beta_1 z} (1 + \Gamma e^{+2j\beta_1 z})$$



At every half-wavelength, all *em* properties repeat

Standing Wave Ratio, s

- ◆ Measures the amount of reflections, the more reflections, the larger the standing wave that is formed.
- ◆ The ratio of $|E_1|_{max}$ to $|E_1|_{min}$

$$S = \frac{|E_1|_{max}}{|E_1|_{min}} = \frac{|H_1|_{max}}{|H_1|_{min}} = \frac{1 + |\Gamma|}{1 - |\Gamma|}$$

or

$$|\Gamma| = \frac{s - 1}{s + 1}$$

Ideally
 $s=1$ (0 dB)
No reflections

PE 10.9

- The plane wave $E=50 \sin (\omega t - 5x) \mathbf{a}_y$ V/m in a lossless medium ($\mu=4\mu_0, \epsilon=\epsilon_0$) encounters a lossy medium ($\mu=\mu_0, \epsilon=4\epsilon_0, \sigma=0.1$ mhos/m) normal to the x -axis at $x=0$. Find

←Answers:

- $\Gamma = 0.8186 \exp(j171^\circ)$
- $\tau = 0.23 \exp(j33.56^\circ)$
- $S = 10.03$
- $E_r = 40.93 \sin (\omega t + 5x + 171^\circ) \mathbf{y}$
- $E_t = 11.5 e^{-6.02x} \sin (\omega t - 7.83x + 33.6^\circ) \mathbf{y}$ V/m
- <http://www.walter-fendt.de/ph14e/stwaverefl.htm>

Ex. Antenna Radome

A 10GHz aircraft radar uses a narrow-beam scanning antenna mounted on a gimbal behind a dielectric radome.

- Even though the radome shape is far from planar, it is approximately planar over the narrow extent of the radar beam.
- If the radome material is a lossless dielectric with $\mu_r=1$ and $\epsilon_r=9$, choose its thickness d such that the radome appears transparent to the radar beam.
- Mechanical integrity requires d to be greater than 2.3 cm.



Antenna with radome



Antenna with no radome

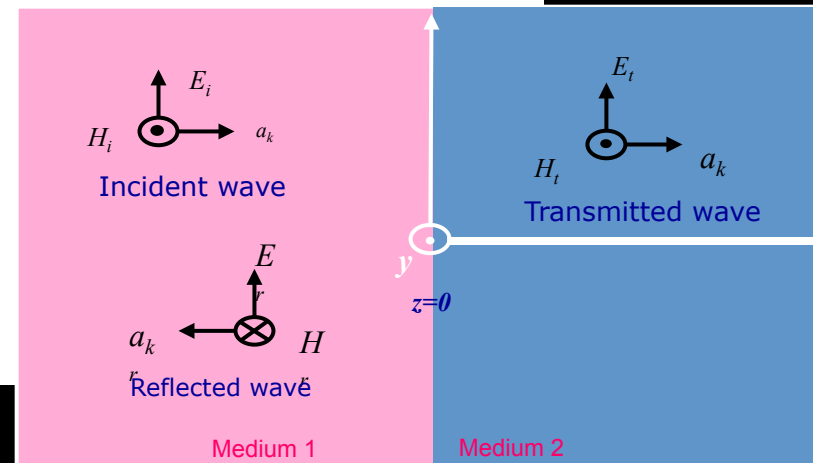
Answer :

$$\lambda/2 = .5\text{cm}, d = 2.5\text{cm}$$

Power Flow in Medium 1

- The net average power density flowing in medium 1

$$\begin{aligned}
 P_{ave1}(z) &= \frac{1}{2} \text{Re}[E_1 \times H_1^*] \\
 &= \frac{1}{2} \text{Re} \left[\hat{x} E_{io} \left(e^{-j\beta_1 z} + \Gamma e^{j\beta_1 z} \right) \times \hat{y} \frac{E_{io}^*}{\eta_1} \left(e^{j\beta_1 z} + \Gamma^* e^{-j\beta_1 z} \right) \right] \\
 &= \hat{z} \frac{|E_{io}|^2}{2\eta_1} \left(1 - |\Gamma|^2 \right) \\
 &= P_{ave}^i + P_{ave}^r
 \end{aligned}$$

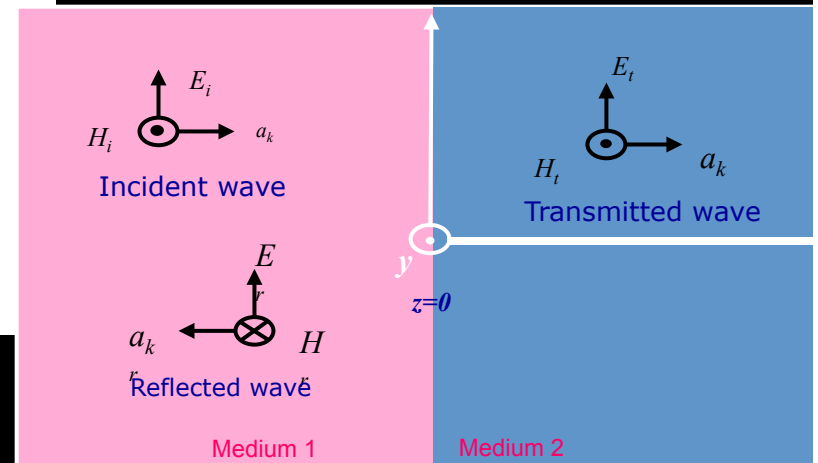


Power Flow in Transmitted wave

- The net average power density flowing in medium 2

$$\begin{aligned}
 P_{ave2}(z) &= \frac{1}{2} \operatorname{Re}[E_2 \times H_2^*] \\
 &= \frac{1}{2} \operatorname{Re} \left[\hat{x} \tau E_{io} e^{-j\beta_2 z} \times \hat{y} \tau^* \frac{E_{io}^*}{\eta_2} e^{j\beta_2 z} \right] \\
 &= \hat{z} |\tau|^2 \frac{|E_{io}|^2}{2\eta_2}
 \end{aligned}$$

where $\hat{a}_k = \hat{z}$



Power in Lossy Media

$$P_{ave1}(z) = \hat{z} \frac{|E_o^i|^2}{2\eta_{c1}} \left(e^{-2\alpha_1 z} - |\Gamma|^2 e^{2\alpha_1 z} \right)$$

$$P_{ave2}(z) = \hat{z} |\tau|^2 \frac{|E_o^i|^2}{2} e^{-2\alpha_2 z} \operatorname{Re} \left(\frac{1}{\eta_{c2}^*} \right)$$

where

$$\Gamma = \frac{\sqrt{\epsilon_{c1}} - \sqrt{\epsilon_{c2}}}{\sqrt{\epsilon_{c1}} + \sqrt{\epsilon_{c2}}} \quad \text{and} \quad \epsilon_{c2} = \epsilon_2 - j \frac{\sigma_2}{\omega_2}$$

quiz

The plane wave $\mathbf{E} = 30 \cos(\omega t - z)\mathbf{a}_x$ V/m in air normally hits a lossless medium ($\mu = \mu_0$, $\epsilon = 4\epsilon_0$) at $z = 0$. (a) Find Γ , τ , and s .

We will look at...

I. Normal incidence

Wave arrives at 90° from the surface

- Standing waves



II. Oblique incidence (lossless)

Wave arrives at an angle

- Snell's Law and Critical angle
- Parallel or Perpendicular
- Brewster angle

